Development of Empirical Models From Process Data

- In some situations it is not feasible to develop a theoretical (physically-based model) due to:
 - 1. Lack of information
 - 2. Model complexity
 - 3. Engineering effort required.
- An attractive alternative: Develop an empirical dynamic model from input-output data.
 - Advantage: less effort is required
 - *Disadvantage:* the model is only valid (at best) for the range of data used in its development.
 - i.e., empirical models usually don't extrapolate very well.

Simple Linear Regression: Steady-State Model

• As an illustrative example, consider a simple linear model between an output variable y and input variable u,

$$y = \beta_1 + \beta_2 u + \varepsilon$$

where β_1 and β_2 are the unknown model parameters to be estimated and ϵ is a random error.

• Predictions of y can be made from the regression model,

$$\hat{y} = \hat{\beta}_1 + \hat{\beta}_2 u \tag{7-3}$$

where $\hat{\beta}_1$ and $\hat{\beta}_2$ denote the estimated values of β_1 and β_2 , and \hat{y} denotes the predicted value of y.

• Let Y denote the measured value of y. Each pair of (u_i, Y_i) observations satisfies:

$$Y_i = \beta_1 + \beta_2 u_i + \varepsilon_i \tag{7-1}$$

The Least Squares Approach

• The least squares method is widely used to calculate the values of β_1 and β_2 that minimize the sum of the squares of the errors S for an arbitrary number of data points, N:

$$S = \sum_{i=1}^{N} \varepsilon_1^2 = \sum_{i=1}^{N} (Y_i - \beta_1 - \beta_2 u_i)^2$$
 (7-2)

• Replace the unknown values of β_1 and β_2 in (7-2) by their estimates. Then using (7-3), S can be written as:

$$S = \sum_{i=1}^{N} e_i^2$$

where the *i*-th residual, e_i , is defined as,

$$e_i \square Y_i - \hat{y}_i \tag{7-4}$$

The Least Squares Approach (continued)

• The least squares solution that minimizes the sum of squared errors, S, is given by:

$$\hat{\beta}_{1} = \frac{S_{uu}S_{y} - S_{uy}S_{u}}{NS_{uu} - (S_{u})^{2}}$$
(7-5)

$$\hat{\beta}_2 = \frac{NS_{uy} - S_u S_y}{NS_{uu} - (S_u)^2}$$
 (7-6)

where:

$$S_u \underline{\Delta} \sum_{i=1}^N u_i$$
 $S_{uu} \underline{\Delta} \sum_{i=1}^N u_i^2$ $S_y \underline{\Delta} \sum_{i=1}^N Y_i$ $S_{uy} \underline{\Delta} \sum_{i=1}^N u_i Y_i$

Extensions of the Least Squares Approach

- Least squares estimation can be extended to more general models with:
 - 1. More than one input or output variable.
 - 2. Functionals of the input variables *u*, such as polynomials and exponentials, as long as the unknown parameters appear linearly.
- A general nonlinear steady-state model which is linear in the parameters has the form,

$$y = \sum_{j=1}^{p} \beta_j X_j + \varepsilon \tag{7-7}$$

where each X_i is a nonlinear function of u.

The sum of the squares function analogous to (7-2) is

$$S = \sum_{i=1}^{N} \left(Y_i - \sum_{j=1}^{p} \beta_j X_{ij} \right)^2$$
 (7-8)

which can be written as,

$$S = (\mathbf{Y} - \mathbf{X} \mathbf{\beta})^{T} (\mathbf{Y} - \mathbf{X} \mathbf{\beta})$$
 (7-9)

where the superscript T denotes the matrix transpose and:

$$\mathbf{Y} = \begin{bmatrix} Y_1 \\ \vdots \\ Y_n \end{bmatrix} \quad \mathbf{\beta} = \begin{bmatrix} \beta_1 \\ \vdots \\ \beta_p \end{bmatrix}$$

$$\boldsymbol{X} = \begin{bmatrix} X_{11} & X_{12} & \cdots & X_{1p} \\ X_{21} & X_{22} & \cdots & X_{2p} \\ \vdots & \vdots & & \vdots \\ X_{n1} & X_{n2} & \cdots & X_{np} \end{bmatrix}$$

The least squares estimates $\hat{\beta}$ is given by,

$$\hat{\boldsymbol{\beta}} = \left(\boldsymbol{X}^T \boldsymbol{X}\right)^{-1} \boldsymbol{X}^T \boldsymbol{Y} \tag{7-10}$$

providing that matrix X^TX is nonsingular so that its inverse exists. Note that the matrix X is comprised of functions of u_j ; for example, if:

$$y = \beta_1 + \beta_2 u + \beta_3 u^2 + \varepsilon$$

This model is in the form of (7-7) if $X_1 = 1$, $X_2 = u$, and $X_3 = u^2$.

Fitting First and Second-Order Models Using Step Tests

- Simple transfer function models can be obtained graphically from step response data.
- A plot of the output response of a process to a step change in input is sometimes referred to as a *process reaction curve*.
- If the process of interest can be approximated by a first- or second-order linear model, the model parameters can be obtained by inspection of the process reaction curve.
- The response of a first-order model, $Y(s)/U(s)=K/(\tau s+1)$, to a step change of magnitude M is:

$$y(t) = KM(1 - e^{-t/\tau})$$
 (5-18)

• The initial slope is given by:

$$\frac{d}{dt} \left(\frac{y}{KM} \right)_{t=0} = \frac{1}{\tau} \tag{7-15}$$

• The gain can be calculated from the steady-state changes in *u* and *y*:

$$K = \frac{\Delta y}{\Delta u} = \frac{\Delta y}{M}$$

where Δy is the steady-state change in y.

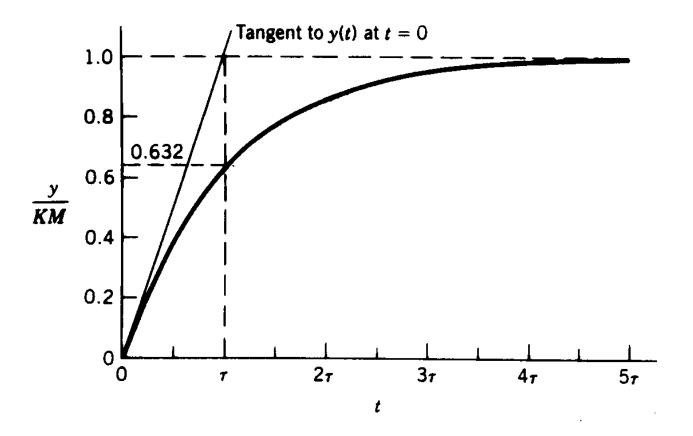


Figure 7.3 Step response of a first-order system and graphical constructions used to estimate the time constant, τ .

First-Order Plus Time Delay Model

$$G(s) = \frac{Ke^{-\theta}s}{\tau s + 1}$$

For this FOPTD model, we note the following characteristics of its step response:

- 1. The response attains 63.2% of its final response at time, $t = \tau + \theta$.
- 2. The line drawn tangent to the response at maximum slope $(t = \theta)$ intersects the y/KM=1 line at $(t = \tau + \theta)$.
- 3. The step response is essentially complete at $t=5\tau$. In other words, the settling time is $t_s=5\tau$.

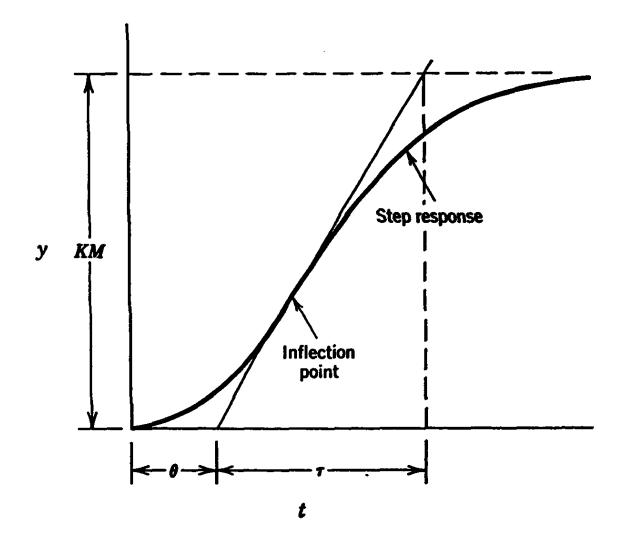


Figure 7.5 Graphical analysis of the process reaction curve to obtain parameters of a first-order plus time delay model.

There are two generally accepted graphical techniques for determining model parameters τ , θ , and K.

Method 1: Slope-intercept method

First, a slope is drawn through the inflection point of the process reaction curve in Fig. 7.5. Then τ and θ are determined by inspection.

Alternatively, τ can be found from the time that the normalized response is 63.2% complete or from determination of the settling time, t_s . Then set $\tau = t_s/5$.

Method 2. Sundaresan and Krishnaswamy's Method

This method avoids use of the point of inflection construction entirely to estimate the time delay.

Sundaresan and Krishnaswamy's Method

- They proposed that two times, t_1 and t_2 , be estimated from a step response curve, corresponding to the 35.3% and 85.3% response times, respectively.
- The time delay and time constant are then estimated from the following equations:

$$\theta = 1.3t_1 - 0.29t_2$$

$$\tau = 0.67(t_2 - t_1)$$
(7-19)

• These values of θ and τ approximately minimize the difference between the measured response and the model, based on a correlation for many data sets.

Estimating Second-order Model Parameters Using Graphical Analysis

- In general, a better approximation to an experimental step response can be obtained by fitting a second-order model to the data.
- Figure 7.6 shows the range of shapes that can occur for the step response model,

$$G(s) = \frac{K}{(\tau_1 s + 1)(\tau_2 s + 1)}$$
 (5-39)

- Figure 7.6 includes two limiting cases: $\tau_2/\tau_1=0$, where the system becomes first order, and $\tau_2/\tau_1=1$, the critically damped case.
- The larger of the two time constants, τ_1 , is called the dominant time constant.

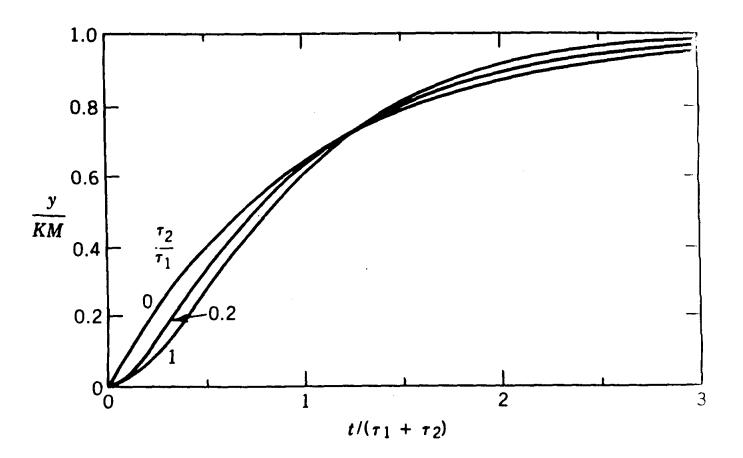


Figure 7.6 Step response for several overdamped second-order systems.

Smith's Method

Assumed model:

$$G(s) = \frac{Ke^{-\theta s}}{\tau^2 s^2 + 2\zeta \tau s + 1}$$

• Procedure:

- 1. Determine t_{20} and t_{60} from the step response.
- 2. Find ζ and t_{60}/τ from Fig. 7.7.
- 3. Find t_{60}/τ from Fig. 7.7 and then calculate τ (since t_{60} is known).

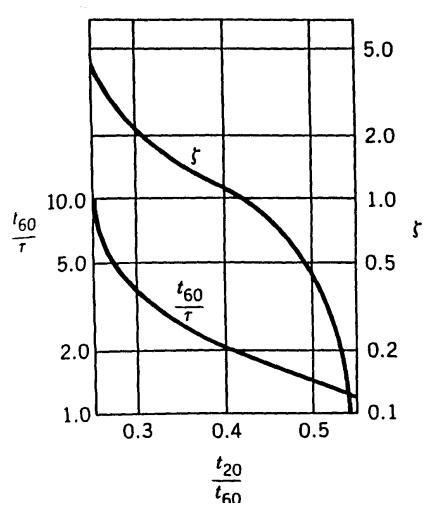


Figure 7.7. Smith's method: relationship of ζ and τ to t_{20} and t_{60} .

Fitting an Integrator Model to Step Response Data

In Chapter 5 we considered the response of a first-order process to a step change in input of magnitude *M*:

$$y_1(t) = KM(1 - e^{-t/\tau})$$
 (5-18)

For short times, $t < \tau$, the exponential term can be approximated by

$$e^{-t/\tau} \approx 1 - \frac{t}{\tau}$$

so that the approximate response is:

$$y_1(t) \approx KM \left[1 - \left(1 - \frac{t}{\tau} \right) \right] = \frac{KM}{\tau} t$$
 (7-22)

is virtually indistinguishable from the step response of the integrating element

$$G_2(s) = \frac{K_2}{s} \tag{7-23}$$

In the time domain, the step response of an integrator is

$$y_2(t) = K_2 M t \tag{7-24}$$

Hence an approximate way of modeling a first-order process is to find the single parameter

$$K_2 = \frac{K}{\tau} \tag{7-25}$$

that matches the early ramp-like response to a step change in input.

If the original process transfer function contains a time delay (cf. Eq. 7-16), the approximate short-term response to a step input of magnitude M would be

$$y(t) = \frac{KM}{t} (t - \theta) S(t - \theta)$$

where $S(t-\theta)$ denotes a delayed unit step function that starts at $t=\theta$.

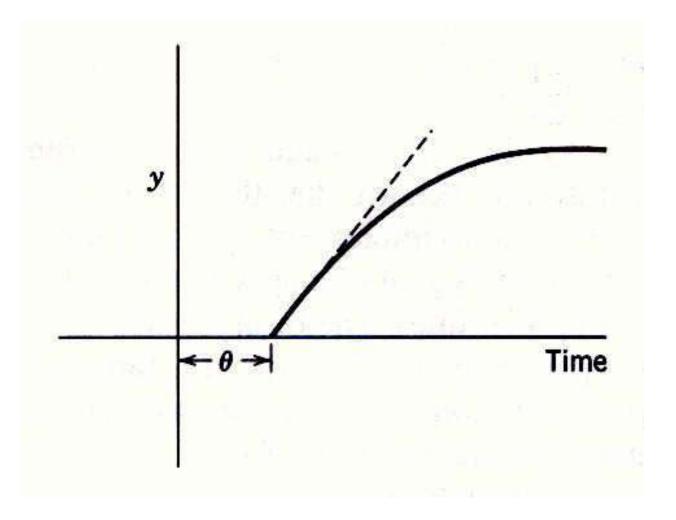


Figure 7.10. Comparison of step responses for a FOPTD model (solid line) and the approximate integrator plus time delay model (dashed line).

Development of Discrete-Time Dynamic Models

- A digital computer by its very nature deals internally with discrete-time data or numerical values of functions at equally spaced intervals determined by the sampling period.
- Thus, discrete-time models such as *difference equations* are widely used in computer control applications.
- One way a continuous-time dynamic model can be converted to discrete-time form is by employing a finite difference approximation.
- Consider a nonlinear differential equation,

$$\frac{dy(t)}{dt} = f(y,u) \tag{7-26}$$

where y is the output variable and u is the input variable.

- This equation can be numerically integrated (though with some error) by introducing a finite difference approximation for the derivative.
- For example, the first-order, backward difference approximation to the derivative at $t = k\Delta t$ is

$$\frac{dy}{dt} \cong \frac{y(k) - y(k-1)}{\Delta t} \tag{7-27}$$

where Δt is the integration interval specified by the user and y(k) denotes the value of y(t) at $t = k\Delta t$. Substituting Eq. 7-26 into (7-27) and evaluating f(y, u) at the previous values of y and u (i.e., y(k-1) and u(k-1)) gives:

$$\frac{y(k) - y(k-1)}{\Delta t} \cong f(y(k-1), u(k-1)) \tag{7-28}$$

$$y(k) = y(k-1) + \Delta t f(y(k-1), u(k-1))$$
 (7-29)

Second-Order Difference Equation Models

- Parameters in a discrete-time model can be estimated directly from input-output data based on linear regression.
- This approach is an example of *system identification* (Ljung, 1999).
- As a specific example, consider the second-order difference equation in (7-36). It can be used to predict y(k) from data available at time $(k-1)\Delta t$ and $(k-2)\Delta t$.

$$y(k) = a_1 y(k-1) + a_2 y(k-2) + b_1 u(k-1) + b_2 u(k-2)$$
 (7-36)

• In developing a discrete-time model, model parameters a_1 , a_2 , b_1 , and b_2 are considered to be unknown.

• This model can be expressed in the standard form of Eq. 7-7,

$$y = \sum_{j=1}^{p} \beta_j X_j + \varepsilon \tag{7-7}$$

by defining:

$$\beta_1 \square a_1, \quad \beta_2 \square a_2, \quad \beta_3 \square b_1, \quad \beta_4 \square b_2$$
 $X_1 \square y(k-1), \quad X_2 \square y(k-2),$
 $X_3 \square u(k-1), \quad X_4 \square u(k-2)$

• The parameters are estimated by minimizing a least squares error criterion:

$$S = \sum_{i=1}^{N} \left(Y_i - \sum_{j=1}^{p} \beta_j X_{ij} \right)^2$$
 (7-8)

Equivalently, S can be expressed as,

$$S = (\mathbf{Y} - \mathbf{X} \mathbf{\beta})^{T} (\mathbf{Y} - \mathbf{X} \mathbf{\beta})$$
 (7-9)

where the superscript T denotes the matrix transpose and:

$$\mathbf{Y} = \begin{bmatrix} Y_1 \\ \vdots \\ Y_n \end{bmatrix} \quad \mathbf{\beta} = \begin{bmatrix} \beta_1 \\ \vdots \\ \beta_p \end{bmatrix}$$

The least squares solution of (7-9) is:

$$\hat{\boldsymbol{\beta}} = \left(\boldsymbol{X}^T \boldsymbol{X}\right)^{-1} \boldsymbol{X}^T \boldsymbol{Y} \tag{7-10}$$